# General Purpose Motion Control ICs 

## Technical Data

## Features

- Low Power CMOS
- PDIP and PLCC Versions Available
- Enhanced Version of the HCTL-1000
- DC, DC Brushless, and Step Motor Control
- Position and Velocity Control
- Programmable Digital Filter and Commutator
- 8-Bit Parallel, and PWM Motor Command Ports
- TTL Compatible
- SYNC Pin for Coordinating Multiple HCTL-1100 ICs
- 100 kHz to 2 MHz Operation
- Encoder Input Port


## Description

The HCTL-1100 series is a high performance, general purpose motion control IC, fabricated in HP CMOS technology. It frees the host processor for other tasks by performing all the time-intensive functions of digital motion control. The programmability of all control parameters provides maximum flexibility and quick

HCTL-1100 Series

## Pinouts


design of control systems with a minimum number of components. In addition to the HCTL-1100, the complete control system consists of a host processor to specify commands, an amplifier, and a motor with an incremental encoder (such as the HP HEDS5XXX, -6XXX, -9XXX series). No analog compensation or velocity feedback is necessary.

## Applications

Typical applications for the HCTL-1100 include printers, medical instruments, material handling machines, and industrial automation.

## HCTL-1100 vs. HCTL-1000

The HCTL-1100 is designed to replace the HCTL-1000. Some differences exist, and some enhancements have been added.


System Block Diagram

Comparison of HCTL-1100 and HCTL-1000

| Description | HCTL-1100 | HCTL-1000 |
| :--- | :--- | :--- |
| Max. Supply Current | 30 mA | 180 mA |
| Max. Power Dissipation | 165 mW | 950 mW |
| Max. Tri-State Output <br> Leakage Current | 150 nA | $10 \mu \mathrm{~A}$ |
| Operating Frequency | $100 \mathrm{kHz}-2 \mathrm{MHz}$ | $1 \mathrm{MHz}-2 \mathrm{MHz}$ |
| Operating Temperature <br> Range | $-20^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Synchronize 2 or More ICs | Yes | - |
| Preset Actual Position <br> Registers | Yes | - |
| Read Flag Register | Yes | - |
| Limit and Stop Pins | Must be pulled <br> up to $\mathrm{V}_{\mathrm{DD}}$ if <br> not used. | Can be left <br> floating if not <br> used. |
| Hard Reset | Required | Recommended |
| PLCC Package Available | Yes | - |

## Package Dimensions



44 PIN PLASTIC LEADED CHIP CARRIER PACKAGE

## Theory of Operation

The HCTL-1100 is a general purpose motor controller which provides position and velocity control for DC, DC brushless and stepper motors. The internal block diagram of the HCTL-1100 is shown in Figure 1. The HCTL1100 receives its input commands from a host processor and position feedback from an incremental encoder with quadrature output. An 8-bit bi-directional multiplexed address/data bus interfaces the HCTL-1100 to the host processor. The encoder
feedback is decoded into quadrature counts and a 24-bit counter keeps track of position. The HCTL-1100 executes any one of four control algorithms selected by the user. The four control modes are:

- Position Control
- Proportional Velocity Control
- Trapezoidal Profile Control for point to point moves
- Integral Velocity Control with continuous velocity profiling using linear acceleration

The resident Position Profile Generator calculates the necessary profiles for Trapezoidal Profile Control and Integral Velocity Control. The HCTL-1100 compares the desired position (or velocity) to the actual position (or velocity) to compute compensated motor commands using a programmable digital filter $\mathrm{D}(\mathrm{z})$. The motor command is externally available at the Motor Command port as an 8-bit byte and at the PWM port as a Pulse Width Modulated (PWM) signal.

The HCTL-1100 has the capability of providing electronic commutation for DC brushless and stepper motors. Using the encoder position information, the motor phases are enabled in the correct sequence. The commutator is fully programmable to encompass most motor/encoder combinations. In addition, phase overlap and phase advance can be programmed to improve torque ripple and high speed performance. The HCTL-1100 contains a number of flags including two externally available flags, Profile and Initialization, which allow the user to see or check the status of the controller. It also has two emergency inputs, Limit and Stop, which allow operation of the HCTL-1100 to be interrupted under emergency conditions.

The HCTL-1100 controller is a digitally sampled data system. While information from the host processor is accepted asynchronously with respect to the control functions, the motor command is computed on a discrete sample time basis. The sample timer is programmable.


Figure 1. Internal Block Diagram.


Figure 2. Operating Mode Flowchart.

## Electrical Specifications

## Absolute Maximum Ratings

Operating Temperature, $\mathrm{T}_{\mathrm{A}}$ $\qquad$ $-20^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$
Storage Temperature, $\mathrm{T}_{\mathrm{S}}$ $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$
Supply Voltage, $\mathrm{V}_{\mathrm{DD}}$
$\qquad$

Input Voltage, $\mathrm{V}_{\mathrm{IN}}$
$\qquad$ -0.3 V to 7 V


DC Electrical Characteristics
$\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} \pm 5 \% ; \mathrm{T}_{\mathrm{A}}=-20^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$

| Parameter | Symbol | Min. | Typ. | Max. | Units | Test Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Voltage | $\mathrm{V}_{\mathrm{DD}}$ | 4.75 | 5.00 | 5.25 | V |  |
| Supply Current | $\mathrm{I}_{\mathrm{DD}}$ |  | 15 | 30 | mA |  |
| Input Leakage Current | $\mathrm{I}_{\mathrm{IN}}$ |  | 10 | 100 | nA | $\mathrm{V}_{\mathrm{IN}}=0.00$ and 5.25 V |
| Input Pull-Up Current |  |  |  |  |  |  |
| SYNC PIN | $\mathrm{I}_{\mathrm{PU}}$ |  | -40 | $\pm 150$ | $\mu \mathrm{~A}$ | $\mathrm{~V}_{\mathrm{IV}}=0.00 \mathrm{~V}$ |
| Tristate Output Leakage <br> Current | $\mathrm{I}_{\mathrm{OZ}}$ |  | 10 | -150 | nA | Sync, LIMIT, STOP <br> pin \#35 (PDIP) <br> V |
|  |  |  |  |  | VUT <br> pin \#38 (PLCC) |  |
| Input Low Voltage | $\mathrm{V}_{\mathrm{IL}}$ | -0.3 |  | 0.8 | V |  |
| Input High Voltage | $\mathrm{V}_{\mathrm{IH}}$ | 2.0 |  | $\mathrm{~V}_{\mathrm{DD}}$ | V |  |
| Output Low Voltage | $\mathrm{V}_{\mathrm{OL}}$ | -0.3 |  | 0.4 | V | $\mathrm{I}_{\mathrm{OL}}=2.2 \mathrm{~mA}$ |
| Output High Voltage | $\mathrm{V}_{\mathrm{OH}}$ | 2.4 |  | $\mathrm{~V}_{\mathrm{DD}}$ | V | $\mathrm{I}_{\mathrm{OH}}=-200 \mu \mathrm{~A}$ |
| Power Dissipation | $\mathrm{P}_{\mathrm{D}}$ |  | 75 | 165 | mW |  |
| Input Capacitance | $\mathrm{C}_{\mathrm{IN}}$ |  |  | 20 | pF |  |
| Output Capacitance | $\mathrm{C}_{\mathrm{OUT}}$ |  | 100 |  | pF |  |

## AC Electrical Characteristics

$\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} \pm 5 \% ; \mathrm{T}_{\mathrm{A}}=-20^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$; Units $=\mathrm{nsec}$

| $\underset{\#}{\text { ID }}$ | Signal | Symbol | Clock Frequency |  |  |  | Formula* |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | 2 MHz |  | 1 MHz |  |  |  |
|  |  |  | Min. | Max. | Min. | Max. | Min. | Max. |
| 1 | Clock Period (clk) | $\mathrm{t}_{\text {CPER }}$ | 500 |  | 1000 |  |  |  |
| 2 | Pulse Width, Clock High | $\mathrm{t}_{\text {cPWH }}$ | 230 |  | 300 |  |  |  |
| 3 | Pulse Width, Clock Low | $\mathrm{t}_{\text {CPWL }}$ | 200 |  | 200 |  | 200 |  |
| 4 | Clock Rise and Fall Time | $\mathrm{t}_{\mathrm{CR}}$ |  | 50 |  | 50 |  | 50 |
| 5 | Input Pulse Width $\overline{\text { Reset }}$ | $\mathrm{t}_{\text {IRST }}$ | 2500 |  | 5000 |  | 5 clk |  |
| 6 | Input Pulse Width $\overline{\text { Stop }}, \overline{\text { Limit }}$ | $\mathrm{t}_{\text {IP }}$ | 600 |  | 1100 |  | $\begin{gathered} 1 \mathrm{clk} \\ +100 \mathrm{~ns} \end{gathered}$ |  |
| 7 | Input Pulse Width $\overline{\text { Index, }}$, Index | $\mathrm{t}_{\text {IX }}$ | 1600 |  | 3100 |  | $\begin{gathered} 3 \mathrm{clk} \\ +100 \mathrm{~ns} \end{gathered}$ |  |
| 8 | Input Pulse Width CHA, CHB | $\mathrm{t}_{\text {IAB }}$ | 1600 |  | 3100 |  | $\begin{gathered} 3 \mathrm{clk} \\ +100 \mathrm{~ns} \end{gathered}$ |  |
| 9 | Delay CHA to CHB Transition | $\mathrm{t}_{\mathrm{AB}}$ | 600 |  | 1100 |  | $\begin{gathered} 1 \mathrm{clk} \\ +100 \mathrm{~ns} \end{gathered}$ |  |
| 10 | Input Rise/Fall Time CHA, CHB, Index | $\mathrm{t}_{\text {IABR }}$ |  | 450 |  | 900 |  | $\begin{gathered} 900 \text { (clk } \\ <1 \mathrm{MHz}) \end{gathered}$ |
| 11 | Input Rise/Fall Time $\overline{\text { Reset }}, \overline{\text { ALE }}$, CS, OE, Stop, Limit | $\mathrm{t}_{\text {IR }}$ |  | 50 |  | 50 |  | 50 |
| 12 | Input Pulse Width $\overline{\mathrm{ALE}}, \overline{\mathrm{CS}}$ | $\mathrm{t}_{\text {IPW }}$ | 80 |  | 80 |  | 80 |  |
| 13 | Delay Time, $\overline{\text { ALE }}$ Fall to $\overline{\mathrm{CS}}$ Fall | $\mathrm{t}_{\mathrm{AC}}$ | 50 |  | 50 |  | 50 |  |
| 14 | Delay Time, $\overline{\text { ALE }}$ Rise to $\overline{\mathrm{CS}}$ Rise | $\mathrm{t}_{\mathrm{CA}}$ | 50 |  | 50 |  | 50 |  |
| 15 | Address Setup Time Before ALE Rise | $\mathrm{t}_{\text {ASR1 }}$ | 20 |  | 20 |  | 20 |  |
| 16 | Address Setup Time Before $\overline{\mathrm{CS}}$ Fall | $\mathrm{t}_{\text {ASR }}$ | 20 |  | 20 |  | 20 |  |
| 17 | Write Data Setup Time Before CS Rise | $\mathrm{t}_{\text {DSR }}$ | 20 |  | 20 |  | 20 |  |
| 18 | Address/Data Hold Time | $\mathrm{t}_{\mathrm{H}}$ | 20 |  | 20 |  | 20 |  |
| 19 | Setup Time, R/ $\overline{\mathrm{W}}$ Before $\overline{\mathrm{CS}}$ Rise | $\mathrm{t}_{\mathrm{wcs}}$ | 20 |  | 20 |  | 20 |  |
| 20 | Hold Time, $\mathrm{R} / \overline{\mathrm{W}}$ After $\overline{\mathrm{CS}}$ Rise | $\mathrm{t}_{\mathrm{wH}}$ | 20 |  | 20 |  | 20 |  |
| 21 | Delay Time, Write Cycle, CS Rise to ALE Fall | $\mathrm{t}_{\text {CSAL }}$ | 1700 |  | 3400 |  | 3.4 clk |  |
| 22 | Delay Time, Read/Write, $\overline{\mathrm{CS}}$ Rise to CS Fall | $\mathrm{t}_{\mathrm{CSCS}}$ | 1500 |  | 3000 |  | 3 clk |  |
| 23 | Write Cycle, $\overline{\text { ALE Fall to }} \overline{\mathrm{ALE}}$ Fall For Next Write | $\mathrm{t}_{\mathrm{wc}}$ | 1830 |  | 3530 |  | 3.7 clk |  |

AC Electrical Characteristics (continued).

| $\underset{\#}{\text { ID }}$ | Signal | Symbol | Clock Frequency |  |  |  | Formula* |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | 2 MHz |  | 1 MHz |  |  |  |
|  |  |  | Min. | Max. | Min. | Max. | Min. | Max. |
| 24 | Delay Time, $\overline{\mathrm{CS}}$ Rise to $\overline{\mathrm{OE}}$ Fall | $\mathrm{t}_{\text {CSOE }}$ | 1700 |  | 3200 |  | $\begin{aligned} & 3 \mathrm{clk} \\ + & 200 \mathrm{~ns} \end{aligned}$ |  |
| 25 | Delay Time, $\overline{\mathrm{OE}}$ Fall to Data Bus Valid | $\mathrm{t}_{\text {OEDB }}$ | 100 |  | 100 |  | 100 |  |
| 26 | Delay Time, $\overline{\mathrm{CS}}$ Rise to Data Bus Valid | $\mathrm{t}_{\text {CSDB }}$ | 1800 |  | 3300 |  | $\begin{aligned} & 3 \mathrm{clk} \\ + & 300 \mathrm{~ns} \end{aligned}$ |  |
| 27 | Input Pulse Width $\overline{\mathrm{OE}}$ | $\mathrm{t}_{\text {IPWOE }}$ | 100 |  | 100 |  | 100 |  |
| 28 | Hold Time, Data Held After $\overline{\text { OE Rise }}$ | $\mathrm{t}_{\text {DOEH }}$ | 20 |  | 20 |  | 20 |  |
| 29 | Delay Time, Read Cycle, $\overline{\mathrm{CS}}$ Rise to $\overline{\text { ALE Fall }}$ | $\mathrm{t}_{\text {CSALR }}$ | 1820 |  | 3320 |  | $\begin{aligned} & 3 \mathrm{clk} \\ + & 320 \mathrm{~ns} \end{aligned}$ |  |
| 30 | Read Cycle, $\overline{\text { ALE }}$ Fall to $\overline{\text { ALE }}$ Fall For Next Read | $\mathrm{t}_{\text {RC }}$ | 1950 |  | 3450 |  | $\begin{aligned} & 3 \mathrm{clk} \\ + & 450 \mathrm{~ns} \end{aligned}$ |  |
| 31 | Output Pulse Width, PROF, INIT, Pulse, Sign, PHA-PHD, MC Port | $\mathrm{t}_{\mathrm{OF}}$ | 500 |  | 1000 |  | 1 clk |  |
| 32 | Output Rise/Fall Time, PROF, INIT, Pulse, Sign, PHA-PHD, MC Port | $\mathrm{t}_{\text {OR }}$ | 20 | 150 | 20 | 150 | 20 | 150 |
| 33 | Delay Time, Clock Rise to Output Rise | $\mathrm{t}_{\text {EP }}$ | 20 | 300 | 20 | 300 | 20 | 300 |
| 34 | Delay Time, CS Rising to MC Port Valid | $\mathrm{t}_{\text {CSMC }}$ |  | 1600 |  | 3200 |  | 3.2 clk |
| 35 | Hold Time, ALE High After CS Rise | $\mathrm{t}_{\text {ALH }}$ | 100 |  | 100 |  | 100 |  |
| 36 | Pulse Width, ALE High | $\mathrm{t}_{\text {ALPWH }}$ | 100 |  | 100 |  | 100 |  |
| 37 | Pulse Width, SYNC Low | $\mathrm{t}_{\text {SYNC }}$ | 9000 |  | 18000 |  | 18 clk |  |

[^0]
## HCTL-1100 I/O Timing Diagrams

Input logic level values are the TTL Logic levels $\mathrm{V}_{\mathrm{IL}}=0.8 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{IH}}=2.0 \mathrm{~V}$. Output logic levels are $\mathrm{V}_{\mathrm{OL}}=0.4 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{OH}}=2.4 \mathrm{~V}$.


## HCTL-1100 I/O Timing Diagrams

There are three different timing configurations which can be used to give the user flexibility to interface the HCTL-1100 to most microprocessors. See the I/O interface section for more details.

## $\overline{\text { ALE/CS }}$ NON OVERLAPPED



Read Cycle


## HCTL-1100 I/O Timing Diagrams

## $\overline{\text { ALE }} / \overline{\mathrm{CS}}$ OVERLAPPED

Write Cycle


## HCTL-1100 I/O Timing Diagrams

## $\overline{\text { ALE WITHIN }} \overline{\mathbf{C S}}$



## Pin Descriptions and Functions

Input/Output Pins

|  | Pin Number |  |  |
| :---: | :---: | :---: | :--- |
| Symbol | PDIP | PLCC | Description |

## Input Signals

| Symbol | Pin Number |  | Description |
| :---: | :---: | :---: | :---: |
|  | PDIP | PLCC |  |
| CHA/CHB | 31, 30 | 34, 33 | Channel A, B - Input pins for position feedback from an incremental shaft encoder. Two channels, A and B, 90 degrees out of phase are required. |
| Index | 33 | 36 | Index Pulse - Input from the reference or index pulse of an incremental encoder. Used only in conjunction with the Commutator. Either a low or high true signal can be used with the Index pin. See Timing Diagrams and Encoder Interface section for more detail. |
| R/W | 37 | 41 | Read/Write - Determines direction of data exchange for the I/O port. |
| ALE | 38 | 42 | Address Latch Enable - Enables lower 6 bits of external data bus into internal address latch. |
| CS | 39 | 43 | Chip Select - Performs I/O operation dependent on status of R/W line. For a Write, the external bus data is written into the internal addressed location. For Read, data is read from an internal location into an internal output latch. |
| OE | 40 | 44 | Output Enable - Enables the data in the internal output latch onto the external data bus to complete a Read operation. |
| Limit | 14 | 15 | Limit Switch - An internal flag which when externally set, triggers an unconditional branch to the Initialization/Idle mode before the next control sample is executed. Motor Command is set to zero. Status of the Limit flag is monitored in the Status register. |
| Stop | 15 | 16 | Stop Flag - An internal flag that is externally set. When flag is set during Integral Velocity Control mode, the Motor Command is decelerated to a stop. |
| $\overline{\text { Reset }}$ | 36 | 40 | Reset - A hard reset of internal circuitry and a branch to Reset mode. |
| ExtClk | 34 | 37 | External Clock |
| $\mathrm{V}_{\text {DD }}$ | 11, 35 | 12,38 | Voltage Supply - Both $\mathrm{V}_{\mathrm{DD}}$ pins must be connected to a 5.0 volt supply. |
| GND | 10, 32 | $\begin{aligned} & 1,11, \\ & 23,35 \end{aligned}$ | Circuit Ground |
| SYNC | 1 | 2 | Used to synchronize multiple HCTL-1100 sample timers. |
| NC | - | 17, 39 | Not connected. These pins should be left floating. |

## Output Pins

| Symbol | Pin Number |  | Description |
| :---: | :---: | :---: | :---: |
|  | PDIP | PLCC |  |
| MC0-MC7 | 18-25 | $\begin{aligned} & 20-22 \\ & 24-28 \end{aligned}$ | Motor Command Port - 8-bit output port which contains the digital motor command adjusted for easy bipolar DAC interfacing. MC7 is the most significant bit (MSB). |
| Pulse | 16 | 18 | Pulse - Pulse width modulated signal whose duty cycle is proportional to the Motor Command magnitude. The frequency of the signal is External Clock/ 100 and pulse width is resolved into 100 external clocks. |
| Sign | 17 | 19 | Sign - Gives the sign/direction of the pulse signal. |
| PHA-PHD | 26-29 | 29-32 | Phase A, B, C, D - Phase Enable outputs of the Commutator. |
| Prof | 12 | 13 | Profile Flag - Status flag which indicates that the controller is executing a profiled position move in the Trapezoidal Profile Control mode. |
| Init | 13 | 14 | Initialization/Idle Flag - Status flag which indicates that the controller is in the Initialization/Idle mode. |

## Pin Functionality

## SYNC Pin

The SYNC pin is used to synchronize two or more ICs. It is only valid in the INIT/IDLE mode (see Operating the HCTL-1100). When this pin is pulled low, the internal sample timer is cleared and held to zero. When the level on the pin is returned to high, the internal sample timer instantly starts counting down from the programmed value.

Connecting all $\overline{\text { SYNC }}$ pins together in the system and pulsing the $\overline{\text { SYNC signal from the }}$ host processor will synchronize all controllers.

## Limit Pin

This emergency-flag input is used to disable the control modes of the HCTL-1100. A low level on this input pin causes the internal Limit flag to be set. If this pin is NOT used, it must be pulled up to $\mathrm{V}_{\mathrm{DD}}$. If it is not connected, the pin could float low, and possibly trigger a false emergency condition.

The Limit flag, when set in any control mode, causes the HCTL1100 to go into the Initialization/ Idle mode, clearing the Motor Command and causing an immediate motor shutdown. When the Limit flag is set, none of the three control mode flags (F0, F3, or F5) are cleared as the HCTL-1100 enters the Initialization/Idle mode. The user should be aware that these flags are still set before commanding the HCTL-1100 to re-enter one of the four control modes from Initialization/Idle mode.

In general, the user should clear all control mode flags after the limit pin has been pulled low, then proceed.

## Stop Pin

The Stop flag affects the HCTL1100 only in the Integral Velocity Mode.

When a low level is present on this emergency-flag input, the internal stop flag is set. If this pin
is NOT used, it must be pulled up to $\mathrm{V}_{\mathrm{DD}}$. If it is not connected, the pin could float low, and possibly trigger a false emergency condition.

When the STOP flag is set, the system will come to a decelerated stop and stay in this mode with a command velocity of zero until the Stop flag is cleared and a new command velocity is specified.

## Notes on Limit and Stop Flags

 Stop and Limit flags are set by a low level input at their respective pins. The flags can only be cleared when the input to the corresponding pin goes high, signifying that the emergency condition has been corrected, AND a write to the Status register $(\mathrm{R} 07 \mathrm{H})$ is executed. That is, after the emergency pin has been set and cleared, the flag also must be cleared by writing to R07H. Any word that is written to R 07 H after the emergency pin is set and cleared will clear the emergency flag. The lower four bits of that word will also reconfigure the Status register.
## Encoder Input Pins (CHA, CHB, INDEX) <br> The HCTL-1100 accepts TTL

 compatible outputs from 2 and 3 channel incremental encoders such as the HEDS-5XXX, 6XXX, and 9XXX series encoders.Channels A and B are internally decoded into quadrature counts which increment or decrement the 24 -bit position counter. For example, a 500 -count encoder is decoded into 2000 quadrature counts per revolution. The position counter will be incremented when Channel B leads Channel A. The Index channel is used only for the Commutator and its function is to serve as a reference point for the internal Ring Counter.

The HCTL-1100 employs an internal 3-bit state delay filter to remove any noise spikes from the encoder inputs to the HCTL-1100. This 3-bit state delay filter requires the encoder inputs to remain stable for three consecutive clock rising edges for an encoder pulse to be considered valid by the HCTL-1100's actual position counter (i.e., an encoder pulse must remain at a logic level high or low for three consecutive clock rising edges for the HCTL1100's actual position counter to be incremented or decremented.) The designer should therefore generally avoid creating the encoder pulses of less than 3 clock cycles.

The index signal of an encoder is used in conjunction with the Commutator. It resets the internal ring counter which keeps track of the rotor position so that no cumulative errors are generated.

The Index pin of the HCTL-1100 also has a 3 -bit filter on its input. The Index pin is active low and level transition sensitive. It detects a valid high-to-low transition and qualifies the low input level through the 3 -bit filter. At this point, the Index signal is internally detected by the commutator logic. This type of configuraiton allows an Index or Index signal to be used to generate the reference mark for commutator operation as long as the AC specifications for the Index signal are met.

## Motor Command Port (MC0MC7)

The 8-bit Motor Command port consists of register R08H whose data goes directly to external pins MC0-MC7. MC7 is the most significant bit. R08H can be read and written to, however, it should be written to only during the Initialization/Idle mode. During any of the four Control modes, the controller writes the motor command into R08H.

This topic is further discussed in the "Register Section" under "Motor Command Register R08H".

## Pulse Width Modulation (PWM) Output Port (Pulse, Sign)

The PWM port consists of the Pulse and Sign pins. The PWM port outputs the motor command as a pulse width modulated signal with the correct polarity. This topic is further discussed in the "Register Section" under "PWM Motor Command Register R09H".

## Trapezoid Profile Pin (Prof)

The Trapezoid Profile Pin is internally connected to software flag bit 4 in the Status Register. This flag is also represented by bit 0 in the Flag Register (R00H). See the "Register Section" for more information. Both the Pin and the Flag indicate the status of a trapezoid profile move. When the HCTL-1100 begins a trapezoid move, this flag is set by the controller (a high level appears on the pin), indicating the move is in progress. When the HCTL-1100 finishes the move, this flag is cleared by the controller.

Note that the instant the flag is cleared may not be the same instant the motor stops. The flag indicates the completion of the command profile, not the actual profile. If the motor is stalled during the move, or cannot physically keep up with the move, the flag will be cleared before the move is finished.

## INIT/IDLE Pin (INIT)

This pin indicates that the HCTL1100 is in the INIT/IDLE mode, waiting to begin control. This pin is internally connected to the software flag bit 5 in the Status Register R07H. This flag is also represented by bit 1 in the Flag Register (R00H) (See the "Register Section" for more information).

Commutator Pins (PHA-PHD)
These pins are connected only when using the commutator of the HCTL-1100 to drive a brushless motor or step motor. The four pins can be programmed to energize each winding on a multiphase motor.

## Operation of the HCTL-1100

## Registers

The HCTL-1100 operation is controlled by a bank of 648 -bit registers, 35 of which are user accessible. These registers
contain command and configuration information necessary to properly run the controller chip. The 35 user-accessible registers are listed in Tables 1 and 2. The register number is also the address. A functional block diagram of the HCTL-1100 which
shows the role of the useraccessible registers is also included in Figure 3. The other 29 registers are used by the internal CPU as scratch registers and should not be accessed by the user.


Figure 3. Register Block Diagram.

## Table 1. Register Reference By Mode

| Register |  | Function | Data Type ${ }^{[1]}$ | User Access |
| :---: | :---: | :---: | :---: | :---: |
| Hex | Dec. |  |  |  |
| General Control |  |  |  |  |
| R00H | R00D | Flag Register |  | r/w |
| R05H | R05D | Program Counter | scalar | r/w |
| R07H | R07D | Status Register | - | $\mathrm{r} / \mathrm{w}^{[2]}$ |
| R0FH | R15D | Sample Timer | scalar | r/w |
| R12H | R18D | Read Actual Position MSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R13H | R19D | Read Actual Position | 2's Complement | $\mathrm{r}^{[4]} / \mathrm{w}^{[5]}$ |
| R14H | R20D | Read Actual Position LSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R15H | R21D | Preset Actual Position MSB | 2's Complement | $\mathrm{w}^{[8]}$ |
| R16H | R22D | Preset Actual Position | 2's Complement | $\mathrm{w}^{[8]}$ |
| R17H | R23D | Preset Actual Position LSB | 2's Complement | $\mathrm{w}^{[8]}$ |
| Output Registers |  |  |  |  |
| R07H | R07D | Sign Reversal Inhibit | - | $\mathrm{r} / \mathrm{w}^{[2]}$ |
| R08H | R08D | 8 bit Motor Command | 2's Complement+80H | $\mathrm{r} / \mathrm{w}$ |
| R09H | R09D | PWM Motor Command | 2's Complement | r/w |
| Filter Registers |  |  |  |  |
| R20H | R32D | Filter Zero, A | scalar | r/w |
| R21H | R33D | Filter Pole, B | scalar | r/w |
| R22H | R34D | Gain, K | scalar | r/w |
| Commutator Registers |  |  |  |  |
| R07H | R07D | Status Register | - | $\mathrm{r} / \mathrm{w}^{[2]}$ |
| R18H | R24D | Commutator Ring | scalar ${ }^{[6,7]}$ | r/w |
| R19H | R25D | Velocity Timer | scalar | W |
| R1AH | R26D | X | scalar ${ }^{[6,7]}$ | r/w |
| R1BH | R27D | Y Phase Overlap | scalar ${ }^{[6,7]}$ | r/w |
| R1CH | R28D | Offset | 2's Complement ${ }^{[7]}$ | r/w |
| R1FH | R31D | Max. Phase Advance | scalar ${ }^{[6,7]}$ | r/w |
| Position Control Mode |  |  |  |  |
| R00H | R00D | Flag Register | - | r/w |
| R12H | R18D | Read Actual Position MSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R13H | R19D | Read Actual Position | 2's Complement | $\mathrm{r}^{[4]} / \mathrm{w}^{[5]}$ |
| R14H | R20D | Read Actual Position LSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R0CH | R12D | Command Position MSB | 2's Complement | $\mathrm{r} / \mathrm{w}^{[3]}$ |
| R0DH | R13D | Command Position | 2's Complement | $\mathrm{r} / \mathrm{w}^{[3]}$ |
| R0EH | R14D | Command Position LSB | 2's Complement | $\mathrm{r} / \mathrm{w}^{[3]}$ |

Table 1. (continued).

| Register |  | Function | Data Type | User Access |
| :---: | :---: | :---: | :---: | :---: |
| Hex | Dec. |  |  |  |
| Trapezoid Profile Control Mode |  |  |  |  |
| R00H | R00D | Flag Register | - | r/w |
| R07H | R07D | Status Register | - | $\mathrm{r} / \mathrm{w}^{[2]}$ |
| R12H | R18D | Read Actual Position MSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R13H | R19D | Read Actual Position | 2's Complement | $\mathrm{r}^{[4]} / \mathrm{w}^{[5]}$ |
| R14H | R20D | Read Actual Position LSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R29H | R41D | Final Position LSB | 2's Complement | r/w |
| R2AH | R42D | Final Position | 2's Complement | r/w |
| R2BH | R43D | Final Position MSB | 2's Complement | r/w |
| R26H | R38D | Acceleration LSB | scalar | r/w |
| R27H | R39D | Acceleration MSB | scalar ${ }^{[6]}$ | r/w |
| R28H | R40D | Maximum Velocity | scalar ${ }^{[6]}$ | $\mathrm{r} / \mathrm{w}$ |
| Integral Velocity Mode |  |  |  |  |
| R00H | R00D | Flag Register | - | r/w |
| R12H | R18D | Read Actual Position MSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R13H | R19D | Read Actual Position | 2's Complement | $\mathrm{r}^{[4]} / \mathrm{w}^{[5]}$ |
| R14H | R20D | Read Actual Position LSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R26H | R38D | Acceleration LSB | scalar | r/w |
| R27H | R39D | Acceleration MSB | scalar ${ }^{[6]}$ | r/w |
| R3CH | R60D | Command Velocity | 2's Complement | r/w |
| Proportional Velocity Mode |  |  |  |  |
| R00H | R00D | Flag Register | - | r/w |
| R12H | R18D | Read Actual Position MSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R13H | R19D | Read Actual Position | 2's Complement | $\mathrm{r}^{[4]} / \mathrm{w}^{[5]}$ |
| R14H | R20D | Read Actual Position LSB | 2's Complement | $\mathrm{r}^{[4]}$ |
| R23H | R35D | Command Velocity LSB | 2's Complement | r/w |
| R24H | R36D | Command Velocity MSB | 2's Complement | r/w |
| R34H | R52D | Actual Velocity LSB | 2's Complement | r |
| R35H | R53D | Actual Velocity MSB | 2's Complement | r |

Notes:

1. Consult appropriate section for data format and use.
2. Upper 4 bits are read only.
3. Writing to ROEH (LSB) latches all 24 bits.
4. Reading R14H (LSB) latches data in R12H and R13H.
5. Writing to R13H clears Actual Position Counter to zero.
6. The scalar data is limited to positive numbers ( 00 H to 7 FH ).
7. The commutator registers ( $\mathrm{R} 18 \mathrm{H}, \mathrm{R1CH}, \mathrm{R} 1 \mathrm{FH}$ ) have further limits which are discussed in the Commutator section of this data sheet.
8. Writing to R17H (R23D) latches all 24 bits (only in INIT/IDLE mode).

Table 2. Register Reference Table by Register Number

| Register |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Hex | Dec. | Function |  |  | Data Type |

## Notes:

1. Consult appropriate section for data format and use.
2. Upper 4 bits are read only.
3. Writing to ROEH (LSB) latches all 24 bits.
4. Reading R14H (LSB) latches data in R12H and R13H.
5. Writing to R13H clears Actual Position Counter to zero.
6. The scalar data is limited to positive numbers $(00 \mathrm{H}$ to 7 FH$)$.
7. The commutator registers ( $\mathrm{R} 18 \mathrm{H}, \mathrm{R} 1 \mathrm{CH}, \mathrm{R} 1 \mathrm{FH}$ ) have further limits which are discussed in the Commutator section of this data sheet.
8. Writing to R17H (R23D) latches all 24 bits (only in INIT/IDLE mode).

## Register Descriptions General Control, Output, Filter, and Commutator

## Flag Register (R00H)

The Flag register contains flags F0 through F5. This register is a read/write register. Each flag is set and cleared by writing an 8 -bit data word to R00H. When writing to R 00 H , the upper four bits are ignored by the HCTL-1100, bits $0,1,2$ specify the flag address, and bit 3 specifies whether to set (bit=1) or clear $(b i t=0)$ the addressed flag.

## Flag Descriptions

F0-Trapezoidal Profile Flag - set by the user to execute Trapezoidal Profile Control. The flag is reset by the controller when the move is completed. The status of F0 can be monitored at the Profile pin and in Status register R07H bit 4.

F1-Initialization/Idle Flag - set/ cleared by the HCTL-1100 to indicate execution of the Initialization/Idle mode. The status of F1 can be monitored at the Initialization/Idle pin and in bit 5 of the Status register (R07H). The user should not attempt to set or clear F1.

F2-Unipolar Flag - set/cleared by the user to specify Bipolar (clear) or Unipolar (set) mode for the Motor Command port.

F3-Proportional Velocity Control Flag - set by the user to specify Proportional Velocity control.

F4-Hold Commutator flag - set/ cleared by the user or automatically by the Align mode. When set, this flag inhibits the internal commutator counters to allow open loop stepping of a
motor by using the commutator. (See "Offset register" description in the "Commutator section.")

F5-Integral Velocity Control - set by the user to specify Integral Velocity Control. Also set and cleared by the HCTL-1100 during execution of the Trapezoidal Profile mode. This is transparent to the user except when the Limit flag is set (see "Emergency Flags" section).

## Writing to the Flag Register

When writing to the flag register, only the lower four bits are used. Bit 3 indicates whether to set or clear a certain flag, and bits 0,1 , and 2 indicate the desired flag. The following table shows the bit map of the Flag register:

| Bit Number | Function |
| :---: | :--- |
| $7-4$ | Don't Care |
| 3 | $1=$ set |
|  | $0=$ clear |
| 2 | AD 2 |
| 1 | AD 1 |
| 0 | AD 0 |

The following table outlines the possible writes to the Flag Register:

| Flag | SET | CLEAR |
| :---: | :---: | :---: |
| F0 | 08 H | 00 H |
| F1 | - | - |
| F2 | 0 AH | 02 H |
| F3 | 0 BH | 03 H |
| F4 | 0 CH | 04 H |
| F5 | 0 DH | 05 H |

## Reading the Flag Register

 Reading register R00H returns the status of the flags in bits 0 to 5. For example, if bit 0 is set (logic 1), then flag F0 is set. If bit 4 is set, then flag F4 is set. If bits 0 and 5 are set, then both flags F0 and F5 are set.The following table outlines the Flag Register Read:

| Bit <br> Number | Flag <br> $(\mathbf{1}=$ set $)$ <br> $(\mathbf{0}=$ clear $)$ |
| :---: | :--- |
| $8-6$ | Don't Care |
| 5 | F5 |
| 4 | F4 |
| 3 | F3 |
| 2 | F2 |
| 1 | F1 |
| 0 | F0 |

Notes:

1. A soft reset (writing 00 H to R 05 H ) will not reset the flags in the flag register. A hard reset ( $\overline{\mathrm{RESET}}$ pin low) is required to reset all the flags. The flags can also be reset by writing the proper word to the Flag register as explained above.
2. While in Trapezoid Profile Mode, Flag F0 will be set, and Flag F5 may be set. F5 is used for internal purposes. Both flags will be cleared at the end of the profile.

## Program Counter Register (R05H)

The Program Counter, which is a write-only register, executes the preprogrammed functions of the controller. The program counter is used along with the control flags F0, F3, and F5 in the Flag register $(\mathrm{R} 00 \mathrm{H})$ to change control modes. The user can write any of the following four commands to the Program Counter.

| Value <br> written <br> to R05H | Action |
| :--- | :--- |
| 00 H | Software Reset |
| 01 H | Enter Init/Idle Mode |
| 02 H | Enter Align Mode <br> (only from INIT/ |
| 03 H | IDLE Mode) <br> Enter Control Mode <br> (only from INIT/ <br> IDLE Mode) |

These Commands are discussed in more detail in the "Operating Modes" section.

## Status Register (R07H)

The Status register indicates the status of the HCTL-1100. Each bit decodes into one signal. All 8 bits are user readable and are decoded as shown below. Only the lower 4 bits can be written to by the user to configure the HCTL-1100. To set or clear any of the lower 4 bits, the user writes an 8-bit word to R 07 H . The upper 4 bits are ignored. Each of the lower 4 bits directly sets/clears the corresponding bit of the Status register as shown below. For example, writing XXXX0101 to R07H sets the PWM Sign Reversal Inhibit, sets the Commutator Phase Configuration to " 3 Phase," and sets the Commutator Count Configuration to "full."

## Motor Command Register (R08H)

The 8-bit Motor Command Port consists of register R08H. The register is connected to external pins MC0-MC7. MC7 is the most significant bit. R08H can be read and written to; however, it should be written to only in the Initialization/Idle mode. During any of the four control modes, the HCTL-1100 writes values to register R08H.

The Motor Command Port operates in two modes, bipolar and unipolar, when under control of internal software. Bipolar mode allows the full range of values in R08H ( -128 D to +127 D ). The data written to the Motor Command Port by the control

Table 3. Status Register

| Status Bit | Function |
| :---: | :--- |
| 0 | PWM Sign Reversal Inhibit <br> $0=$ off <br> $1=$ on |
| 1 | Commutator Phase Configuration <br> $0=3$ phase <br> $1=4$ phase |
| 2 | Commutator Count Configuration <br> $0=$ quadrature <br> $1=$ full |
| 3 | Should always be set to 0 |
| 4 | Trapezoidal Profile Flag F0 <br> $1=$ in Profile Control |
| 5 | Initialization/Idle Flag F1 <br> $1=$ in Initialization/Idle Mode |
| 6 | Stop Flag <br> $0=$ set (Stop triggered) <br> $1=$ cleared (no Stop) |
| 7 | Limit Flag <br> $0=$ set (Limit triggered) <br> $1=$ cleared (no Limit) |

## PWM Motor Command Register (R09H)

The PWM port outputs the motor command as a pulse width modulated signal with the correct sign of polarity. The PWM port consists of the Pulse and Sign pins and R09H.

The PWM signal at the Pulse pin has a frequency of External Clock/100 and the duty cycle is
resolved into the 100 clocks. (For example, a 2 MHz clock gives a 20 KHz PWM frequency.)

The Sign pin gives the polarity of the command. Low output on Sign pin is positive polarity.

The 2's-complement contents of R09H determine the duty cycle and polarity of the PWM command. For example, D8H
(-40D) gives a 40\% duty cycle signal at the Pulse pin and forces the Sign pin high. Data outside the $64 \mathrm{H}(+100 \mathrm{D})$ to $9 \mathrm{CH}(-$ 100D) linear range gives 100\% duty cycle. R09H can be read and written to. However, the user should only write to R 09 H when the controller is in the Initialization/Idle mode.
Figure 5 shows the PWM output versus the internal motor command.


Figure 4. Motor Command Port Output.

When any Control mode is being executed, the unadjusted internal 2's-complement motor command is written to R 09 H . Because of the hardware limit on the linear range $(64 \mathrm{H}$ to $9 \mathrm{CH}, \pm 100 \mathrm{D})$, the PWM port saturates sooner than the 8bit Motor Command port ( 00 H to FFH, +127D to -128D). When the internal motor command saturates above 8 bits, the PWM port is saturated to the full
$\pm 100 \%$ duty cycle level. Figure 5 shows the actual values inside the PWM port. Note that the Unipolar flag, F2, does not affect the PWM port.

For commutation of brushless motors with the PWM port, only use the Pulse pin from the PWM port as the commutator already contains sign information. (See Figure 9.)

The PWM port has an option that can be used with H-bridge type amplifiers. The option is Sign Reversal Inhibit, which inhibits the Pulse output for one PWM period after a sign polarity reversal. This allows one pair of transistors to turn off before others are turned on and thereby avoids a short across the power supply. Bit 0 in the Status register (R07H) controls the Sign Reversal


Figure 5. PWM Port Output.


Figure 6. Sign Reversal Inhibit.

Inhibit option. Figure 6 shows the output of the PWM port when Bit 0 is set.

## Actual Position Registers

Read, Clear: R12H,R13H,R14H
Preset : R15H,R16H,R17H
The Actual Position Register is accessed by two sets of registers in the HCTL-1100. When reading the Actual Position from the HCTL-1100, the host processor will read Registers R12H(MSB), R13H, and R14H(LSB). When presetting the Actual Position Register, the processor will write to Registers R15H(MSB), R16H, and R17H(LSB).

When reading the Actual Position registers, the order should be R14H, R13H, R12H. These registers are latched, such that, when reading Register R14H, all three bytes will be latched so that count data does not change while reading three separate bytes.

When presetting the Actual Position Register, write to R15H and R16H first. When R17H is written to, all three bytes are simultaneously loaded into the Actual Position Register.

Note that presetting the Actual Position Registers is only allowed while the HCTL-1100 is in INIT/ IDLE mode.

The Actual Position Registers can be simultaneously cleared at any time by writing any value to

R13H.

## Digital Filter Registers

Zero (A) R20H
Pole (B) R21H
Gain (K) R22H

All control modes use some part of the programmable digital filter $\mathrm{D}(\mathrm{z})$ to compensate for closed loop system stability. The compensation $\mathrm{D}(\mathrm{z})$ has the form:

$$
\begin{equation*}
\mathrm{D}(\mathrm{z})=\frac{\mathrm{K}\left(\mathrm{z}-\frac{\mathrm{A}}{256}\right)}{4\left(\mathrm{z}+\frac{\mathrm{B}}{256}\right)} \tag{1}
\end{equation*}
$$

where:
$\mathrm{z}=$ the digital domain operator
$\mathrm{K}=$ digital filter gain (R22H)
$\mathrm{A}=$ digital filter zero (R20H)
$B=$ digital filter pole (R21H)
The compensation is a first-order lead filter which in combination with the Sample Timer T (R0FH) affects the dynamic step response and stability of the control system. The Sample Timer, T, determines the rate at which the control algorithm gets executed. All parameters, A, B, K, and T, are 8 -bit scalars that can be changed by the user any time.

As shown in equations [2] and [3], the digital filter uses previously sampled data to calculate $\mathrm{D}(\mathrm{z})$. This old internally sampled data is cleared when the Initialization/Idle mode is executed.

In Position Control, Integral Velocity Control, and Trapezoidal Profile Control the digital filter is implemented in the time domain as shown below:

$$
\begin{align*}
\mathrm{MC}_{\mathrm{n}}= & (\mathrm{K} / 4)\left(\mathrm{X}_{\mathrm{n}}\right)- \\
& {\left[(\mathrm{A} / 256)(\mathrm{K} / 4)\left(\mathrm{X}_{\mathrm{n}-1}\right)+\right.} \\
& \left.(\mathrm{B} / 256)\left(\mathrm{MC}_{\mathrm{n}-1}\right)\right] \tag{2}
\end{align*}
$$

where:

$$
\begin{aligned}
& \mathrm{n}=\text { current sample time } \\
& \mathrm{n}-1=\text { previous sample time } \\
& \mathrm{MC}_{\mathrm{n}}=\text { Motor Command Output } \\
& \text { at } \mathrm{n} \\
& \mathrm{MC}_{\mathrm{n}-1}=\text { Motor Command } \\
& \text { Output at } \mathrm{n}-1 \\
& \mathrm{X}_{\mathrm{n}}=\text { (Command Position }- \\
& \text { Actual Position) at } \mathrm{n} \\
& \mathrm{X}_{\mathrm{n}-1}=(\text { Command Position }- \\
& \text { Actual Position) at } \mathrm{n}-1
\end{aligned}
$$

In Proportional Velocity control the digital compensation filter is implemented in the time domain as:

$$
\begin{equation*}
\mathrm{MC}_{\mathrm{n}}=(\mathrm{K} / 4)\left(\mathrm{Y}_{\mathrm{n}}\right) \tag{3}
\end{equation*}
$$

where:

$$
\begin{gathered}
\mathrm{Y}_{\mathrm{n}}=(\text { Command Velocity }- \\
\text { Actual Velocity }) \text { at } \mathrm{n}
\end{gathered}
$$

For more information on system sampling times, bandwidth, and stability, please consult HewlettPackard Application Note 1032, Design of the HCTL-1000's
Digital Filter Parameters by the Combination Method.

Sample Timer Register (ROFH)
The contents of this register set the sampling period of the HCTL1100. The sampling period is:

$$
\begin{aligned}
\mathrm{t}= & 16(\mathrm{~T}+1)(1 / \text { frequency of the } \\
& \text { external clock })
\end{aligned}
$$

where:
$\mathrm{T}=$ contents of register R0FH

The Sample Timer has a limit on the minimum allowable sample time depending on the control mode being executed. The limits are given in Table 4 below.

The minimum value limits are to make sure the internal programs have enough time to complete proper execution.

The maximum value of $\mathrm{T}(\mathrm{ROFH})$ is FFH (255D). With a 2 MHz clock, the sample time can vary from $64 \mu \mathrm{sec}$ to $2048 \mu \mathrm{sec}$. With a 1 MHz clock, the sample time can vary from $128 \mu \mathrm{sec}$ to 4096 $\mu \mathrm{sec}$.

Digital closed-loop systems with slow sampling times have lower stability and a lower bandwidth than similar systems with faster sampling times. To keep the system stability and bandwidth as high as possible the HCTL-1100 should typically be programmed with the fastest sampling time
possible. This rule of thumb must be balanced by the needs of the velocity range to be controlled. Velocities are specified to the HCTL-1100 in terms of quadrature encoder counts per sample time. The faster the sampling time, the higher the slowest possible speed.

## Hardware Description

The Sample Timer consists of a buffer and a decrement counter. Each time the counter reaches $00 H$, the Sampler Timer Value T (value written to R0FH) is loaded from the buffer into the counter, which immediately begins to decrement from T.

## Writing to the Sample Timer Register

Data written to R0FH will be latched into the internal buffer and used by the counter after it completes the present sample time cycle by decrementing to 00 H . The next sample time will use the newly written data.

## Reading the Sample Timer Register

Reading R0FH gives the values directly from the decrementing counter. Therefore, the data read from R0FH will have a value anywhere between T and 00 H , depending where in the sample time cycle the counter is.

## Table 4.

| Control Mode | R0FH Contents <br> Minimum Limit |
| :--- | :---: |
| Position Control | $07 \mathrm{H}(07 \mathrm{D})$ |
| Proportional Velocity Control | $07 \mathrm{H}(07 \mathrm{D})$ |
| Trapezoidal Profile Control | $0 \mathrm{FH}(15 \mathrm{D})$ |
| Integral Velocity Control | $0 \mathrm{FH}(15 \mathrm{D})$ |

Example -

1. On reset, the value of the timer is pre-set to 40 H .
2. Reading R0FH shows 3EH . . . 2BH . . . 08H . . . 3CH . . .

## Synchronizing Multiple Axes

Synchronizing multiple axes with HCTL-1100s can be achieved by using the $\overline{\text { SYNC }}$ pin as explained in the Pin Discussion section. Some users may not only want to synchronize several HCTL-1100s but also follow custom profiles for each axis. To do this, the user may need to write a new command position or command velocity during each sample time for the duration of the profile. In this case, data written to the HCTL-1100 has to be coordinated with the Sample Timer. This is so that only one command position or velocity is received during any one sample period, and that it is written at the proper time within a sample period.

At the beginning of each sample period, the HCTL-1100 is performing calculations and executions. New command positions and velocities should not be written to the HCTL-1100 during this time. If they are, the calculations may be thrown off and cause unpredictable control.

The user can read the Sample Timer Register to avoid writing too early during a sample period. Since the Sample Timer Register continuously counts down from its programmed value, the user can check if enough time has passed in the sample period to insure the completion of the internal calculations. The length of time needed by the HCTL-1100
to do its calculations is given by the Minimum Limits of R0FH (Sample Timer Register) as shown in Table 4. For Position Control Mode, the user should wait for the Sample Timer to count down 07H from its programmed value before writing the next command position or velocity. If the programmed sample timer value is 39 H , wait until the Sample Timer Register reads 32H. Writing between 32 H and 00 H will make the command information available for the next sample period.

## Commutator

Status Register
(R07H)
Commutator Ring
(R18H)
X Register
(R1AH)
Y Phase Overlap (R1BH)
Offset ( R 1 CH )
Max. Phase Advance
Velocity Timer

Phase advance allows the user to compensate for the frequency characteristics of the motor/ amplifier combination. By advancing the phase enable command (in position), the delay in reaction of the motor/amplifier combination can be offset and higher performance can be achieved.

Phase offset is used to adjust the alignment of the commutator output with the motor torque curves. By correctly aligning the HCTL-1100's commutator output with the motor's torque curves, maximum motor output torque can be achieved.

The inputs to the Commutator are the three encoder signals, Channel A, Channel B, and Index, and the configuration data stored in registers.

The Commutator uses both channels and the index pulse of an incremental encoder. The index pulse of the encoder must be physically aligned to a known torque curve location because it is used as the reference point of the rotor position with respect to the Commutator phase enables.

The index pulse should be permanently aligned during motor encoder assembly to the last motor phase. This is done by energizing the last phase of the motor during assembly and permanently attaching the encoder codewheel to the motor shaft such that the index pulse is active as shown in Figures 7 and 8. Fine tuning of alignment for commutation purposes is done electronically by the Offset register ( R 1 CH ) once the complete control system is set up.

The commutator is a digital state machine that is configured by the user to properly select the phase sequence for electronic commutation of multiphase motors. The Commutator is designed to work with 2,3 , and 4phase motors of various winding configurations and with various encoder counts. Along with providing the correct phase enable sequence, the Commutator provides programmable phase overlap, phase advance, and phase offset.

Phase overlap is used for better torque ripple control. It can also be used to generate unique state sequences which can be further decoded externally to drive more complex amplifiers and motors.


Figure 7. Index Pulse Alignment to Motor Torque Curves.

Each time an index pulse occurs, the internal commutator ring counter is reset to 0 . The ring counter keeps track of the current position of the rotor based on the encoder feedback. When the ring counter is reset to 0 , the Commutator is reset to its origin (last phase going low, Phase A going high) as shown in Figure 10.

The output of the Commutator is available as PHA, PHB, PHC, and PHD. The HCTL-1100's commutator acts as the electrical equivalent of the mechanical brushes in a motor. Therefore, the outputs of the commutator provide only proper phase sequencing for bidirectional operation. The magnitude information is provided to the
motor via the Motor Command and PWM ports. The outputs of the commutator must be combined with the outputs of one of the motor ports to provide proper DC brushless and stepper motor control. Figure 9 shows an example of circuitry which uses the outputs of the commutator with the Pulse output of the PWM port to control a DC brushless or


Figure 8. Codewheel Index Pulse Alignment.


Figure 9. PWM Interface to Brushless DC Motors.


Figure 10. Commutator Configuration.
stepper motor. A similar procedure could be used to combine the commutator outputs PHAPHD with a linear amplifier interface output (Figure 16) to create a linear amplifier system.

The Commutator is programmed by the data in the following registers. Figure 10 shows an example of the relationship between all the parameters.

## Status Register (R07H)

Bit \#1- $0=3$-phase configuration, PHA, PHB, and PHC are active outputs.
$1=4$-phase configuration, PHA - PHD are active outputs.
Bit \#2- $0=$ Rotor position measured in quadrature counts ( 4 x decoding).
$1=$ Rotor position measured in full counts ( 1 count $=1$ codewheel bar and space.)

Bit \#2 only affects the commutator's counting method. This includes the Ring register ( R 18 H ), the X and Y registers (R1AH \& R1BH), the Offset register ( R 1 CH ), the Velocity Timer register (R19H), and the Maximum Advance register (R1FH).

Quadrature counts ( 4 x decoding) are always used by the HCTL1100 as a basis for position, velocity, and acceleration control.

## Ring Register (R18H)

The Ring register is defined as 1 electrical cycle of the commutator which corresponds to 1 torque cycle of the motor. The Ring register is scalar and determines the length of the commutation
cycle measured in full or quadrature counts as set by bit \#2 in the Status register ( R 07 H ). The value of the ring must be limited to the range of 0 to 7 FH .

## X Register (R1AH)

This register contains scalar data which sets the interval during which only one phase is active.

## Y Register (R1BH)

This register contains scalar data which set the interval during which two sequential phases are both active. Y is phase overlap. X and Y must be specified such that:

$$
\mathrm{X}+\mathrm{Y}=\text { Ring } /(\# \text { of phases })[5]
$$

These three parameters define the basic electrical commutation cycle.

## Offset Register (R1CH)

The Offset register contains two's-complement data which determines the relative start of the commutation cycle with respect to the index pulse. Since the index pulse must be physically referenced to the rotor, offset performs fine alignment between the electrical and mechanical torque cycles.

The Hold Commutator flag (F4) in the Status register ( R 07 H ) is used to decouple the internal commutator counters from the encoder input. Flag (F4) can be used in conjunction with the Offset register to allow the user to advance the commutator phases open loop. This technique may be used to create a custom commutator alignment procedure. For example, in Figure 10, case 1, for a three-phase motor where the ring $=9, X=3$, and $Y=0$, the phases can be made to advance open loop by setting the Hold Commutator flag (F4) in the Flag register $(\mathrm{R} 07 \mathrm{H})$. When the values

0,1 , or 2 are written to the Offset register, phase A will be enabled. When the values 3 , 4 or 5 are written to the Offset register, phase B will be enabled. And, when the values 6,7 , or 8 are written to the Offset register, phase C will be enabled. No values larger than the value programmed into the Ring register should be programmed into the Offset register.

## Phase Advance Registers (R19H, R1FH)

The Velocity Timer register and Maximum Advance register linearly increment the phase advance according to the measured speed for rotation up to a set maximum.

The Velocity Timer register (R19H) contains scalar data which determines the amount of phase advance at a given velocity. The phase advance is interpreted in the units set for the Ring counter by bit \#2 in R07H. The velocity is measured in revolutions per second.

$$
\begin{aligned}
& \text { Advance }=\mathrm{N}_{\mathrm{f}} \mathrm{v} \Delta \mathrm{t} \\
& \text { where: } \Delta \mathrm{t}=\frac{16(\mathrm{R} 19 \mathrm{H}+1)}{\mathrm{f} \text { external clk }} \\
& \mathrm{N}_{\mathrm{f}}=\begin{array}{l}
\text { full encoder counts/ } \\
\text { revolution. }
\end{array} \\
& \mathrm{v}=\begin{array}{l}
\text { velocity (revolutions/ } \\
\text { second })
\end{array}
\end{aligned}
$$

The Maximum Advance register (R1FH) contains scalar data which sets the upper limit for phase advance regardless of rotor speed.

Figure 11 shows the relationship between the Phase Advance registers. Note: If the phase advance feature is not used, set both R19H and R1FH to 0 .

## Commutator Constraints and Use

When choosing a three-channel encoder to use with a DC brushless or stepper motor, the user should keep in mind that the number of quadrature encoder counts ( $4 x$ the number of slots in the encoder's codewheel) must be an integer multiple ( $1 \mathrm{x}, 2 \mathrm{x}, 3 \mathrm{x}$, $4 x, 5 x$, etc.) of the number of pole pairs in the DC brushless motor or steps in a stepper motor. To take full advantage of the commutator's overlap feature, the number of quadrature counts should be at least 3 times the number of pole pairs in the DC brushless motor or steps in the stepper motor. For example, a $1.8^{\circ}$, (200 step/revolution) stepper motor should employ at least a 150 slot codewheel $=600$

quadrature counts/revolution $=3$ x 200 steps/revolution).

There are several numerical constraints the user should be aware of to use the Commutator.

The parameters of Ring, X, Y, and Max Advance must be positive numbers $(00 \mathrm{H}$ to 7 FH$)$. Additionally, the following equation must be satisfied:

$$
\begin{aligned}
& (-128 \mathrm{D}) 80 \mathrm{H} \leq 3 / 2 \text { Ring } \\
& \quad+\text { Offset } \pm \text { Max Advance } \\
& \leq 7 \mathrm{FH}(127 \mathrm{D})
\end{aligned}
$$

In order to utilize the greatest flexibility of the Commutator, it must be realized that the Commutator works on a circular ring counter principle, whose range is defined by the Ring register ( R 18 H ). This means that

## 3. Ring register $=\frac{(4)(192) \text { counts } / \text { revolution }}{8 / \text { revolution }}$ <br> $=96$ quadrature counts <br> $=1$ commutation cycle

4. By measuring the motor torque curve in both directions, it is determined that an offset of 3 mechanical degrees, and a phase overlap of 2 mechanical degrees is needed.

$$
\begin{aligned}
\text { Offset } & =3^{\circ} \frac{(4)(192)}{360^{\circ}} \\
& \cong 6 \text { quadrature counts }
\end{aligned}
$$

[^1]To create the 3 mechanical degree offset, the Offset register ( R 1 CH ) could be programmed with either A6H (-90D) or $06 \mathrm{H}(+06 \mathrm{D})$. However, because 06H (+06D) would violate the commutator constraints Equation 8, A6H (-90D) is used.

$$
\begin{gathered}
\mathrm{Y}=\text { overlap }=\frac{\left(2^{\circ}\right)(4)(192)}{360^{\circ}} \cong 4 \\
\mathrm{X}+\mathrm{Y}=96 / 3 \\
\text { Therefore }, \mathrm{X}=28 \\
\mathrm{Y}=4
\end{gathered}
$$

For the purposes of this example, the Velocity Timer and Maximum Advance are set to 0 .

## Operation Flowchart

The HCTL-1100 executes any one of three setup routines or four control modes selected by the user. The three setup routines include:

- Reset
- Initialization/Idle
- Align.

The four control modes available to the user include:

- Position Control
- Proportional Velocity Control
- Trapezoidal Profile Control
- Integral Velocity Control

The HCTL-1100 switches from one mode to another as a result of one of the following three mechanisms:

1. The user writes to the Program Counter.
2. The user sets/clears flags F0, F3, or F5 by writing to the Flag register ( ROOH ).
3. The controller switches automatically when certain initial conditions are provided by the user.


Figure 12. Operation Flowchart.

This section describes the function of each setup routine and control mode and the initial conditions which must be provided by the user to switch from
one mode to another. Figure 12 shows a flowchart of the setup routines and control modes, and shows the commands required to switch from one mode to another.

## Setup Modes

## Hard Reset

Executed by:
-Pulling the RESET pin low (required at power up)

When a hard reset is executed (RESET pin goes low), the following conditions occur:

- All output signal pins are held low except Sign, Data bus, and Motor Command.
- All flags (F0 to F5) are cleared.
- The Pulse pin of the PWM port is set low while the $\overline{\text { Reset }}$ pin is held low. After the $\overline{\text { Reset }} \mathrm{pin}$ is released (goes high) the Pulse pin goes high for one cycle of the external clock driving the HCTL-1100. The Pulse pin then returns to a low output.
- The Motor Command port $(\mathrm{R} 08 \mathrm{H})$ is preset to 80 H (128D).
- The Commutator logic is cleared.
- The I/O control logic is cleared.
- A soft reset is automatically executed.


## Soft Reset

Executed by:

- Writing 00H to R05H, or
- Automatically called after a hard reset

When a soft reset is executed, the following conditions occur:

- The digital filter parameters are preset to $\mathrm{A}(\mathrm{R} 20 \mathrm{H})=\mathrm{E} 5 \mathrm{H}(229 \mathrm{D})$ $\mathrm{B}(\mathrm{R} 21 \mathrm{H})=\mathrm{K}(\mathrm{R} 22 \mathrm{H})=40 \mathrm{H}$ (64D)
- The Sample Timer (R0FH) is preset to 40 H (64D).
- The Status register (R07H) is cleared.
- The Actual Position Counters (R12H, R13H, R14H) are cleared to 0 .

From Reset mode, the HCTL-1100 goes automatically to
Initialization/Idle mode.

## Initialization/Idle

Executed by:

- Writing 01 H to R 05 H , or
- Automatically executed after a hard reset, soft reset, or
- Limit pin goes low.

The Initialization/Idle mode is entered either automatically from Reset, by writing 01 H to the Program Counter (R05H) under any conditions, or pulling the Limit pin low.

In the Initialization/Idle mode, the following occur:

- The Initialization/Idle flag (F1) is set.
- The PWM port R09H is set to 00 H (zero command).
- The Motor Command port ( R 08 H ) is set to 80 H (128D) (zero command).
- Previously sampled data stored in the digital filter is cleared.

It is at this point that the user should pre-program all the necessary registers needed to execute the desired control mode. The HCTL-1100 stays in this mode (idling) until a new mode command is given.

## Align

Executed by:

- Writing 02H to R05H

The Align mode is executed only when using the commutator feature of the HCTL-1100. This mode automatically aligns multiphase motors to the HCTL1100's internal Commutator.

The Align mode can be entered only from the Initialization/Idle mode by writing 02 H to the Program Counter register (R05H).

Before attempting to enter the Align mode, the user should clear all control mode flags and set both the Command Position registers ( $\mathrm{R} 0 \mathrm{CH}, \mathrm{RODH}$, and ROEH) and the Actual Position registers (R12H, R13H, and R14H) to zero. After the Align mode has been executed, the HCTL-1100 will automatically enter the Position Control mode and go to position zero. By following this procedure, the largest movement in the Align mode will be one torque cycle of the motor.

The Align mode assumes: the encoder index pulse has been physically aligned to the last motor phase during encoder/ motor assembly, the Commutator parameters have been correctly preprogrammed (see the section called Commutator for details), and a hard reset has been executed while the motor is stationary.

The Align mode first disables the Commutator and with open loop control enables the first phase (PHA) and then the last phase (PHC or PHD) to orient the motor on the last phase torque detent. Each phase is energized for 2048 system sampling periods (t). For proper operation, the motor must come to a complete stop during the last phase enable. At this point the Commutator is enabled and commutation is closed loop.

The HCTL-1100 then automatically switches from the Align mode to Position Control mode.

## Control Modes

Control flags F0, F3, and F5 in the Flag register (R00H) determine which control mode is executed. Only one control flag can be set at a time. After one of
these control flags is set, the control modes are entered either automatically from Align or from the Initialization/Idle mode by writing 03 H to the Program Counter (R05H).

Position Control Mode
Flags: F0 Cleared
F3 Cleared
F5 Cleared

Registers Used:
Register Function
R00H R00D Flag Register
R12H R18D Read Actual Position MSB
R13H R19D Read Actual Position
R14H R20D Read Actual Position LSB
R0CH R12D Command Position MSB
R0DH R13D Command Position
R0EH R14D Command Position LSB

Position Control performs point-to-point position moves with no velocity profiling. The user specifies a 24 -bit position
command, which the controller compares to the 24-bit actual position. The position error is calculated, the full digital lead compensation is applied and the motor command is output.

The controller will remain position-locked at a destination until a new position command is given.

The actual and command position data is 24-bit two's-complement data stored in six 8-bit registers. Position is measured in encoder quadrature counts.

The command position resides in R0CH (MSB), R0DH, R0EH (LSB). Writing to ROEH latches all 24 bits at once for the control algorithm. Therefore, the command position is written in the sequence $\mathrm{R} 0 \mathrm{CH}, \mathrm{RODH}$ and ROEH. The command registers can be read in any desired order.

The actual position resides in R12H (MSB), R13H, and R14H (LSB). Reading R14H latches the upper two bytes into an internal buffer. Therefore, Actual Position
registers are read in the order of $\mathrm{R} 14 \mathrm{H}, \mathrm{R} 13 \mathrm{H}$, and R12H for correct instantaneous position data.

The largest position move possible in Position Control mode is 7 FFFFFH (8,388,607D) quadrature encoder counts.

Proportional Velocity Mode
Flags: F0 Cleared F3 Set
F5 Cleared

Registers Used:

| Register | Function |
| :---: | :--- |
| R00H R00D | Flag Register |
| R23H R35D | Command |
|  | Velocity LSB |
| R24H R36D | Command |
|  | Velocity MSB |
| R34H R52D | Actual Velocity |
|  | LSB |
| R35H R53D | Actual Velocity |
|  | MSB |

Proportional Velocity Control performs control of motor speed using only the gain factor, K, for compensation. The dynamic pole and zero lead compensation are not used. (See the "Digital Filter" section of this data sheet.)

```
Example Code to Program Position Moves
    { Begin }
            Hard Reset { HCTL-1100 goes into INIT/IDLE Mode }
            Initialize Filter, Timer, Command Position Registers
            Write 03H to Register R05H
            { HCTL-1100 is now in Position Mode }
            Write Desired Command Position to Command Position Registers
            { Controller Moves to new position }
            Continue writing in new Command Positions
{ end }
```

The command and actual velocity are 16-bit two's-complement words.

The command velocity resides in registers R 24 H (MSB) and R23H (LSB). These registers are unlatched which means that the command velocity will change to a new velocity as soon as the value in either R23H or R24H is changed. The registers can be read or written to in any order.


The units of velocity are quadrature counts/sample time. To convert from rpm to quadrature counts/sample time, use the formula shown below:
$\mathrm{Vq}=(\mathrm{Vr})(\mathrm{N})(\mathrm{t})(0.01667 / \mathrm{rpm}-\mathrm{sec})[9$

Where:
$\mathrm{Vq}=$ velocity in quadrature
counts/sample time
$\mathrm{Vr}=$ velocity in rpm
$\mathrm{N}=4$ times the number of slots
in the codewheel (i.e., quadrature counts).
$\mathrm{t}=$ The HCTL-1100 sample time in seconds. (See the section on the HCTL-1100's Sample Timer register).

Because the Command Velocity registers ( R 24 H and R 23 H ) are internally interpreted by the HCTL-1100 as 12 bits of integer and 4 bits of fraction, the host processor must multiply the desired command velocity (in quadrature counts/sample time) by 16 before programming it into the HCTL-1100's Command Velocity registers.

The actual velocity is computed only in this algorithm and stored in scratch registers R35H (MSB) and R34H (LSB). There is no fractional component in the actual velocity registers and they can be read in any order.

The controller tracks the command velocity continuously until new mode command is given. The system behavior after a new velocity command is governed only by the system dynamics until a steady state velocity is reached.

## Integral Velocity Mode <br> Flags: F0 Cleared <br> F3 Cleared <br> F5 Set to begin move

Registers Used:

| Register | Function |
| :--- | :--- |
| R00H R00D | Flag Register |
| R26H R38D | Acceleration LSB |
| R27H R39D | Acceleration MSB |
| R3CH R60D | Command |
|  | Velocity |

Integral Velocity Control performs continuous velocity profiling which is specified by a command velocity and command acceleration. Figure 13 shows the capability of this control algorithm.

The user can change velocity and acceleration any time to continuously profile velocity in time. Once the specified velocity is reached, the HCTL-1100 will maintain that velocity until a new command is specified. Changes between actual velocities occur at the presently specified linear acceleration.

The command velocity is an 8-bit two's-complement word stored in R3CH. The units of velocity are

```
Example Code for Programming Proportional Velocity Mode
    { Begin }
            Hard Reset { HCTL-1100 goes into INIT/IDLE Mode }
            Initialize Filter, Timer, Command Position Registers
            Write 03H to Register R05H
                    { HCTL-1100 is now in Position Mode }
                            Write Desired Command Velocity (if needed)
            Set Flag F3 {Proportional Velocity Move Begins}
            { System ramps to Command Velocity }
            Continue writing new Command Velocities
        {end}
```

quadrature counts/sample time.
The conversion from rpm to quadrature counts/sample time is shown in equation 9 . The Command Velocity register (R3CH) contains only integer data and has no fractional component.

While the overall range of the velocity command is 8 bits, two'scomplement, the difference between any two sequential commands cannot be greater than 7 bits in magnitude (i.e., 127 decimal). For example, when the HCTL-1100 is executing a command velocity of 40 H (+64D), the next velocity command must fall in the range of $7 \mathrm{FH}(+127 \mathrm{D})$, the maximum command range, $\mathrm{C} 1 \mathrm{H}(-63 \mathrm{D})$, the largest allowed difference.

The command acceleration is a 16-bit scalar word stored in R27H and R 26 H . The upper byte ( R 27 H ) is the integer part and the lower byte ( R 26 H ) is the fractional part provided for resolution. The integer part has a range of 00 H to 7 FH . The contents of R26H are internally divided by 256 to produce the

fractional resolution.
The units of acceleration are quadrature counts/sample time squared.

To convert from rpm/sec to quadrature counts/[sample time] ${ }^{2}$, use the formula shown below:
$\mathrm{Aq}=(\mathrm{Ar})(\mathrm{N})\left(\mathrm{t}^{2}\right)(0.01667 / \mathrm{rpm}-$ $\mathrm{sec})$

Where:
$\mathrm{Aq}=$ Acceleration in quadrature counts/[sample time] ${ }^{2}$
$\mathrm{Ar}=$ Acceleration in rpm/sec
$\mathrm{N}=4$ times the number of slots in the codewheel (i.e., quadrature counts)
$\mathrm{t}=$ The HCTL-1100 sample time in seconds. (See the section on the HCTL-1100's Sample Timer register).

Because the Command Acceleration registers (R27H and R26H) are internally interpreted by the HCTL-1100 as 8 bits of integer and 8 bits of fraction, the host processor must multiply the desired command acceleration (in quadrature counts/[sample time] ${ }^{2}$ ) by 256 before programming it into the HCTL-1100's Command Acceleration registers.

Internally, the controller performs velocity profiling through position control.

Each sample time, the internal profile generator uses the information which the user has programmed into the Command

Velocity register (R3CH) and the Command Acceleration registers ( R 27 H and R 26 H ) to determine the value which will be automatically loaded into the Command Position registers (ROCH, RODH, and ROEH). After the new command position has been generated, the difference between the value in the Actual Position registers (R12-R13H, and R14H) and the new value in the Command Position registers is calculated as the new position error. This new position error is used by the full digital compensation filter to compute a new motor command output by this sample time. The register block in Figure 3 further shows how the internal profile generator works in Integral Velocity mode. In control theory terms, integral compensation has been added and therefore, this system has zero steadystate error.

Although Integral Velocity Control mode has the advantage over Proportional Velocity mode of zero steady state velocity error, its disadvantage is that the closed
loop stability is more difficult to achieve. In Integral Velocity Control mode the system is actually a position control system and therefore the complete dynamic compensation $D(z)$ is used.

If the external Stop flag F6 is set during this mode signalling an emergency situation, the controller automatically decelerates to zero velocity at the presently specified acceleration factor and stays in this condition until the flag is cleared. The user then can specify new velocity profiling data.

| Trapezoid Profile Mode |  |
| :--- | :--- | :--- |
| Flags: | F0 Set to begin move |
|  | F3 Cleared |
| F5 Cleared |  |
|  |  |
| Registers Used: |  |
| Register | Function |
| R00H R00D | Flag Register |
| R07H R07D | Status Register |
| R12H R18D | Read Actual |
|  | Position MSB |
| R13H R19D | Read Actual |
|  | Position |
| R14H R20D | Read Actual |
|  | Position LSB |
| R29H R41D | Final Position |
|  | LSB |
| R2AH R42D | Final Position |
| R2BH R43D | Final Position |
|  | MSB |
| R26H R38D | Acceleration LSB |
| R27H R39D | Acceleration MSB |
| R28H R40D | Maximum |
|  | Velocity |

Trapezoid Profile Control performs point-to-point position moves and profiles the velocity trajectory to a trapezoid or triangle. The user specifies only the desired final position, acceleration and maximum velocity. The controller computes the necessary profile to conform to the command data. If maximum velocity is reached before the distance halfway point, the profile will be trapezoidal, otherwise the profile will be triangular. Figure 14 shows the possible trajectories with Trapezoidal Profile Control.

The command data for Trapezoidal Profile Control mode consists of a final position, a command acceleration, and a maximum velocity. The 24-bit,

```
Example Code for Programming Integral Velocity Mode
(Begin)
    Hard Reset {HCTL-1100 goes into INIT/IDLE Mode}
    Initialize Filter, Timer, Command Position Registers
    Write 03H to Register R05H
            {HCTL-1100 is now in Position Mode}
    Write Desired Acceleration (if needed)
    Write Desired Maximum Velocity (if needed)
    Set Flag F5 {Integral Velocity Move Begins}
    {System ramps to Maximum Velocity}
    Continue writing new Accelerations and Velocities
{ end }
```



Figure 14. Trapezoidal Profile Mode.
two's-complement final position is written to registers R2BH, (MSB), R2AH, and R29H (LSB). The 16bit command acceleration resides in registers R27H (MSB) and R26H (LSB). The command acceleration has the same integer and fraction format as discussed in the Integral Velocity Control mode section. The 7 -bit maximum velocity is a scalar value with the range of 00 H to 7 FH ( 0 D to 127D). The maximum velocity has the units of quadrature counts per sample time, and resides in register R 28 H . The command data registers may be read or written to in any order.

The internal profile generator produces a position profile using the present Command Position (ROCH-ROEH) as the starting point and the Final Position ( $\mathrm{R} 2 \mathrm{BH}-\mathrm{R} 29 \mathrm{H}$ ) as the end point.

Once the desired data is entered, the user sets flag F0 in the Flag register $(\mathrm{R} 00 \mathrm{H})$ to commence motion (if the HCTL-1100 is already in Position Control mode).

When the profile generator sends the last position command to the Command Position registers to complete the trapezoidal move, the controller clears flag F0. The HCTL-1100 then automatically goes to Position Control mode with the final position of the trapezoidal move as the command position.

When the HCTL-1100 clears flag F0 it does NOT indicate that the motor and encoder are at the final position NOR that the motor and encoder have stopped. The flag indicates that the command profile has finished. The motor and encoder's true position can only be determined by reading the Actual Position registers. The only way to determine if the motor and encoder have stopped is to read the Actual Position registers at successive intervals.

The status of the Profile flag can be monitored both in the Status register (R07) and at the external Profile pin at any time. While the Profile flag is high NO new
command data should be sent to the controller.

Each sample time, the internal profile generator uses the information which the user has programmed into the Maximum Velocity register (R28H), the Command Acceleration registers (R27H and R26H), and the Final Position registers (R2BH, R2AH, and R 29 H ) to determine the value which will be automatically loaded into the Command Position registers (ROEH, RODH, and R0CH). After the new command position has been generated, the difference between the value in the Actual Position registers ( $\mathrm{R} 12 \mathrm{H}, \mathrm{R} 13 \mathrm{H}$, and R 14 H ) and the new value in the Command Position registers is calculated as the new position error. This new position error is used by the full digital compensation filter to compute a new motor command output for the sample time. (The register block diagram in Figure 3 further shows how the internal profile generator works in Trapezoidal Profile mode.)

```
Example Code for Programming Trapezoid Moves
    { Begin }
        Hard Reset { HCTL-1100 goes into INIT/IDLE Mode }
        Inititalize Filter, Timer, Command Position Registers
    Write 03H to Register R05H
        { HCTL-1100 is now in Position Mode }
    { Profile #1}
    Write Desired Acceleration
    Write Desired Maximum Velocity
    Write Final Position
    Set Flag FO {Trapezoid Move Begins, PROF pin goes high}
    Poll PROF pin until it goes low (Move is complete)
    { Profile #2}
    Write Desired Acceleration
    Write Desired Maximum Velocity
    Write Final Position
    Set Flag FO {Trapezoid Move Begins, PROF pin goes high}
    Poll PROF pin until it goes low (Move is complete)
    { Repeat }
    •
    •
    .
{ end }
```

Applications of the HCTL-1100
Interfacing the HCTL-1100 to Host Processors
The HCTL-1100 looks to the host microprocessor like a bank of 8bit registers to which the host
processor can read and write (i.e., the host processor treats the HCTL-1100 like RAM). The data in these registers controls the operation of the HCTL-1100. The host processor communicates to
the HCTL-1100 over a bidirectional multiplexed 8 -bit data bus. The four I/O control lines. $\overline{\mathrm{ALE}}, \overline{\mathrm{CS}}, \overline{\mathrm{OE}}$, and $\mathrm{R} / \overline{\mathrm{W}}$ execute the data transfers (see Figure 15).

There are three different timing configurations which can be used to give the user greater flexibility to interface the HCTL-1100 to most microprocessors (see Timing diagrams). They are differentiated from one another by the arrangement of the ALE signal with respect to the CS signal. The three timing configurations are listed below.

1. $\overline{\mathrm{ALE}}, \overline{\mathrm{CS}}$ non-overlapped
2. $\overline{\mathrm{ALE}}, \overline{\mathrm{CS}}$ overlapped
3. $\overline{\mathrm{ALE}}$ within $\overline{\mathrm{CS}}$

Any I/O operation starts by asserting the $\overline{\text { ALE }}$ signal which starts sampling the external bus into an internal address latch. Rising $\overline{\text { ALE }}$ or falling $\overline{\mathrm{CS}}$ during ALE stops the sampling into the address latch.

CS low after rising ALE samples the external bus into the data latch. Rising $\overline{\mathrm{CS}}$ stops the sampling into the data latch, and starts the internal synchronous process.

In the case of a write, the data in the data latch is written into the addressed location. In the case of a read, the addressed location is written into an internal output latch. $\overline{\mathrm{OE}}$ low enables the internal output latch onto the external bus. The $\overline{\mathrm{OE}}$ signal and the internal output latch allow the I/O port to be flexible and avoid bus conflicts during read operations.

It is important that the host microprocessor does not attempt to perform too many I/O operations in a single sample time of the HCTL-1100. Each I/O operation interrupts the execution of the HCTL-1100's internal code for 1 clock cycle.

Although extra clock cycles have been allotted in each sample time for I/O operations, the number of extra cycles is reduced as the value programmed into the Sample Timer register (ROFH) is reduced.

Table 5 shows the maximum number of I/O operations allowed under the given conditions.

The number of external clock cycles available for I/O operations in any of the four control modes can be increased by increasing the value in the Sample Timer register (R0FH).

For every unit increase in the Sample Timer register (R0FH) above the minimums shown in Table 5 the user may perform 16 additional I/O operations per sample time.

## Interfacing the HCTL-1100 to Amplifiers and Motors

The Motor Command port is the ideal interface to an 8-bit DAC, configured for bipolar output. The
data written to the 8-bit Motor Command port by the control algorithms is the internally computed 2's-complement motor command with an 80 H offset added. This allows direct interfacing to a DAC. Figure 16 shows a typical DAC interface to the HCTL-1100. An inexpensive DAC, such as MC1408 or equivalent, has its digital inputs directly connected to the Motor Command port. The DAC produces an output current which is converted to a voltage by an operational amplifier. $\mathrm{R}_{\mathrm{o}}$ and $\mathrm{R}_{\mathrm{G}}$ control the analog offset and gain. The circuit is easily adjusted for +5 V to -5 V operation by first writing 80 H to R 08 H and adjusting $R_{\mathrm{o}}$ for 0 V output. Then FFH is written to R 08 H and $\mathrm{R}_{\mathrm{G}}$ is adjusted until the output is 5 V . Note that 00H in R08H corresponds to -5 V out.

Figure 17 shows an example of how to interface the HCTL-1100 to an H-bridge amplifier. An H bridge amplifier allows bipolar motor operation with a unipolar power supply.

## Table 5. Maximum Number of I/O Allowed

| Sample Timer <br> Register Value | Operating Mode | Maximum Number <br> of I/O Operations <br> Allowed per Sample |
| :---: | :---: | :---: |
| 07 H (07D) | Position Control or <br> Prop. Vel. Control | 5 |
| OFH (15D) | Position Control or <br> Prop. Vel. Control <br> Trapezoidal Prof. <br> or Integral <br> Vel. Control | 133 |



Figure 15. I/O Port Block Diagram.


Figure 16. Linear Amplifier Interface.


Figure 17. H-Bridge Amplifier Interface.

## Additional Information From Hewlett-Packard

Application notes and Application briefs regarding the HCTL-1100 is are from the Hewlett-Packard Motion Control Factory. Please contact your local HP sales representative for more information.

- M003 - Z80 Interface to the HCTL-1100
- M005 - Sample Timer and Digital Filter
- M009 - List of Board Level Vendors Using HCTL-1100
- M010 - HCTL-1100 Trouble Shooting Guide
- M012 - Commutator Port in the HCTL-1100
- M015 - Interfacing the HCTL1100 to the 8051
- M016-8051/HCTL-1100 Stand Alone Controller with RS232 Port
- M018 - The Effects of HighFrequency Noise on the HCTL1100
- M021 - Interfacing the HCTL1100 to 68 HC 11 .
- M024 - Using the HCTL-1100 with DC Brush Motors.
- M025 - Using the HCTL-1100 with DC Brushless Motors.
- M026 - Using the HCTL-1100 with Stepper Motors.


## Ordering Information

HCTL-1100: 40 Pin DIP
Package
HCTL-1100\#PLC: 44 Pin PLCC
Package


[^0]:    *General formula for determining AC characteristics for other clock frequencies (clk), between 100 kHz and 2 MHz .

[^1]:    Figure 11. Phase Advance vs. Motor Velocity.

