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ADJUSTABLE LED DRIVER

Check for Samples: TL4242-Q1

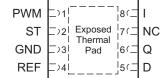
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FEATURES

- Qualified for Automotive Applications
- AEC-Q100 Test Guidance With the Following Results:
 - Device Temperature Grade 2: -40°C to 105°C Ambient Operating Temperature Range for QFN package
 - Device Temperature Grade 1: -40°C to 125°C Ambient Operating Temperature Range for SFM package
 - Device HBM ESD Classification Level H1C
 - Device CDM ESD Classification Level C3B
- Adjustable Constant Current up to 500 mA (±5%)
- Wide Input-Voltage Range up to 42 V
- Low Dropout Voltage
- Open-Load Detection
- Overtemperature Protection

DESCRIPTION

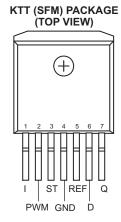




Short-Circuit Proof

Reverse-Polarity Proof

NC – No internal connection



The TL4242-Q1 is an integrated adjustable constant-current source, driving loads up to 500 mA. One can adjust the output current level through an external resistor. The device design is for supplying high-power LEDs (for example, OSRAM Dragon LA W57B) under the severe conditions of automotive applications, resulting in constant brightness and extended LED lifetime. The device comes in the DRJ (QFN) package. Protection circuits prevent damage to the device in case of overload, short circuit, reverse polarity, and overheat. The device provides the connected LEDs protection against reverse polarity as well as excess voltages up to 45 V.

The integrated PWM input of the TL4242-Q1 permits LED brightness regulation by pulse-width modulation (PWM). The high input impedance of the PWM input, allows operating the LED driver as a protected high-side switch.

An external shunt resistor in the ground path of the connected LEDs senses the LED current. A regulation loop holds the voltage drop at the shunt resistor at a constant level of 177 mV (typical). The selection of the shunt resistance, R_{REF} , sets the constant-current level. Calculate the typical output current using the equation:

$$I_{Q,typ} = \frac{V_{REF}}{R_{REF}}$$

where V_{REF} is the reference voltage (typically 177 mV) (see *Reference Electrical Characteristics*). The equation applies for R_{REF} = 0.39 Ω to 10 Ω .

The output current is shown as a function of the reference resistance in . With the PWM input, One can regulate the LED brightness through the duty cycle. Also, PWM = L sets the TL4242-Q1 in sleep mode, resulting in a very low current consumption of < 1 μ A (typical). The high impedance of the PWM input (see) permits the use of the PWM pin as an enable input.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

TL4242-Q1



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ORDERING INFORMATION⁽¹⁾

T _A	ORDERABLE PART NUMBER ⁽²⁾	TOP-SIDE MARKING
–40°C to 105°C	TL4242TDRJRQ1	4242T
-40°C to 125°C	TL4242QKTTRQ1	TL4242Q

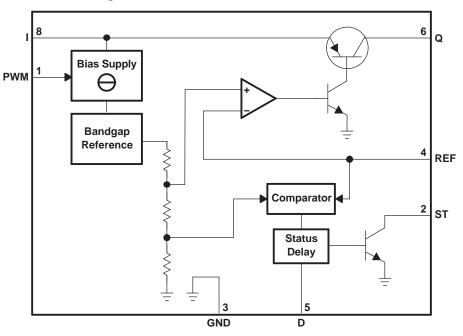
(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

(2) Package drawings, thermal data, and symbolization are available at www.ti.com/packaging.

PIN FUNCTIONS

	N	0.	DESCRIPTION
NAME	DRJ	КТТ	DESCRIPTION
D	5	6	Status delay. To set status reaction delay, connect to GND with a capacitor. For no delay, leave open.
GND	3	4	Ground
NC	7	N/A	No internal connection
PWM	1	2	Pulse-width modulation input. If not used, connect to I.
Q	6	7	Output
REF	4	5	Reference input. Connect to a shunt resistor.
ST	2	3	Status output. Open-collector output. Connect to an external pullup resistor ($R_{PULLUP} \ge 4.7 \text{ k}\Omega$).
I	8	1	Input. Connect directly to GND as close as possible to the device with a 100-nF ceramic capacitor.
Thermal pad	-	-	Solder the thermal pad directly to the PCB. Connect to ground or leave floating

Figure 1. FUNCTIONAL BLOCK DIAGRAM



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ABSOLUTE MAXIMUM RATINGS⁽¹⁾

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
V_{CC}	Supply voltage range ⁽²⁾		-42	45	V
		D	-0.3	7	V
VI	Input voltage range	PWM	-40	40	V
		REF	-1	16	V
V _O Outpu		Q	-1	41	V
	output voltage range	ST	-0.3	40	V
		PWM		±1	mA
I _O	Output current range	REF		±2	mA
		ST		±5	mA
TJ	Virtual-junction temperature range		-40	150	°C
T _{stg}	Storage temperature range		-50	150	°C
		Human-body model (HBM) AEC-Q100 Classification Level H1C		1500	
ESD	Electrostatic discharge rating	Machine model (MM) AEC-Q100 Classification Level M3		200	V
		Charged-device model (CDM) AEC-Q100 Classification Level C3B		1000	

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values are with respect to the network ground terminal.

THERMAL INFORMATION

	THERMAL METRIC ⁽¹⁾	TL42	TL4242-Q1		
			KTT (7 PINS)	UNIT	
θ_{JA}	Junction-to-ambient thermal resistance	39	31.6		
θ_{JCtop}	Junction-to-case (top) thermal resistance	31.5	34.7		
θ_{JB}	Junction-to-board thermal resistance	15.5	8.2	°C/W	
ΨJT	Junction-to-top characterization parameter	0.3	0.7	C/VV	
Ψ_{JB}	Junction-to-board characterization parameter	15.6	8.2		
θ_{JCbot}	Junction-to-case (bottom) thermal resistance	1.8	0.7		

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, SPRA953. **RECOMMENDED OPERATING CONDITIONS**

		MIN	MAX	UNIT
V _{CC}	Supply voltage	4.5	42	V
V_{ST}	Status (ST) output voltage		16	V
V _{PWM}	PWM voltage	0	40	V
CD	Status delay (D) capacitance	0	2.2	μF
R _{REF}	Reference (REF) resistor	0	10	Ω
T _A	Operating free-air temperature, QFN	-40	105	°C
T _A	Operating free-air temperature, SFM	-40	125	°C

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OVERALL DEVICE ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, $V_I = 13.5 \text{ V}$, $R_{REF} = 0.47 \Omega$, $V_{PWM,H}$, $T_A = -40^{\circ}\text{C}$ to 105°C (QFN), $T_A = -40^{\circ}\text{C}$ to 125°C (SFM), all voltages with respect to ground (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{qL}	Supply current	V _Q = 6.6 V		12	22	mA
I_{qOFF}	Supply current, off mode	$PWM = L, T_J < 85^{\circ}C$		0.1	2	μA

OUTPUT ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, $V_I = 13.5 \text{ V}$, $R_{REF} = 0.47 \Omega$, $V_{PWM,H}$, $T_A = -40^{\circ}\text{C}$ to 105°C (QFN), $T_A = -40^{\circ}\text{C}$ to 125°C (SFM), all voltages with respect to ground (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Ι _Q		$V_Q - V_{REF}^{(1)} = 6.6 V$	357	376	395	
	Output ourset	$V_Q - V_{REF} = 6.6 \text{ V}, \text{ R}_{REF} = 1 \Omega$	168	177	185	
		$V_Q - V_{REF} = 6.6 \text{ V}, \text{R}_{REF} = 0.39 \Omega$	431	454	476	- 1
		$V_Q - V_{REF} = 5.4$ V to 7.8 V, $V_I = 9$ V to 16 V	357	376	395	
I _{Qmax}	Output current limit	$R_{REF} = 0 \Omega$		600		mA
V _{dr}	Drop voltage	I _Q = 300 mA		0.35	0.7	V

(1) $V_Q - V_{REF}$ equals the forward voltage sum of the connected LEDs (see).

PWM INPUT ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, $V_I = 13.5 \text{ V}$, $R_{REF} = 0.47 \Omega$, $V_{PWM,H}$, $T_A = -40^{\circ}C$ to $105^{\circ}C$ (QFN), $T_A = -40^{\circ}C$ to $125^{\circ}C$ (SFM), all voltages with respect to ground (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{PWM,H}	High-level PWM voltage		2.6			V
V _{PWM,L}	Low-level PWM voltage				0.7	V
I _{PWM,H}	High-level PWM input current	V _{PWM} = 5 V		220	500	μA
I _{PWM,L}	Low-level PWM input current	V _{PWM} = 0 V	-1		1	μA
t _{PWM,ON}	Delay time, turnon	70% of I _{Qnom} , see Figure 7	0	15	40	μs
t _{PWM,OFF}	Delay time, turnoff	30% of I _{Qnom} , see Figure 7	0	15	40	μs

REFERENCE (REF) ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, $V_I = 13.5 \text{ V}$, $R_{REF} = 0.47 \Omega$, $V_{PWM,H}$, $T_A = -40^{\circ}$ C to 105°C (QFN), $T_A = -40^{\circ}$ C to 125°C (SFM), all voltages with respect to ground (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{REF}	Reference voltage	$R_{REF} = 0.39 \Omega$ to 1 Ω	168	177	185	mV
I _{REF}	Reference input current	V _{REF} = 180 mV	-1	0.1	1	μA

STRUMENTS

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STATUS OUTPUT (ST) ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, $V_I = 13.5 \text{ V}$, $R_{REF} = 0.47 \Omega$, $V_{PWM,H}$, $T_A = -40^{\circ}C$ to 105°C (QFN), $T_A = -40^{\circ}C$ to 125°C (SFM), all voltages with respect to ground (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IQL}	Lower status-switching threshold	ST = L	15	25		mV
V _{IQH}	Upper status-switching threshold	ST = H		30	40	mV
V_{STL}	Low-level status voltage	I _{ST} = 1.5 mA			0.4	V
I _{STLK}	Leakage current	V _{ST} = 5 V			5	μA

STATUS DELAY (D) ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, $V_I = 13.5 \text{ V}$, $R_{REF} = 0.47 \Omega$, $V_{PWM,H}$, $T_A = -40^{\circ}$ C to 105°C (QFN), $T_A = -40^{\circ}$ C to 125°C (SFM), all voltages with respect to ground (unless otherwise noted)

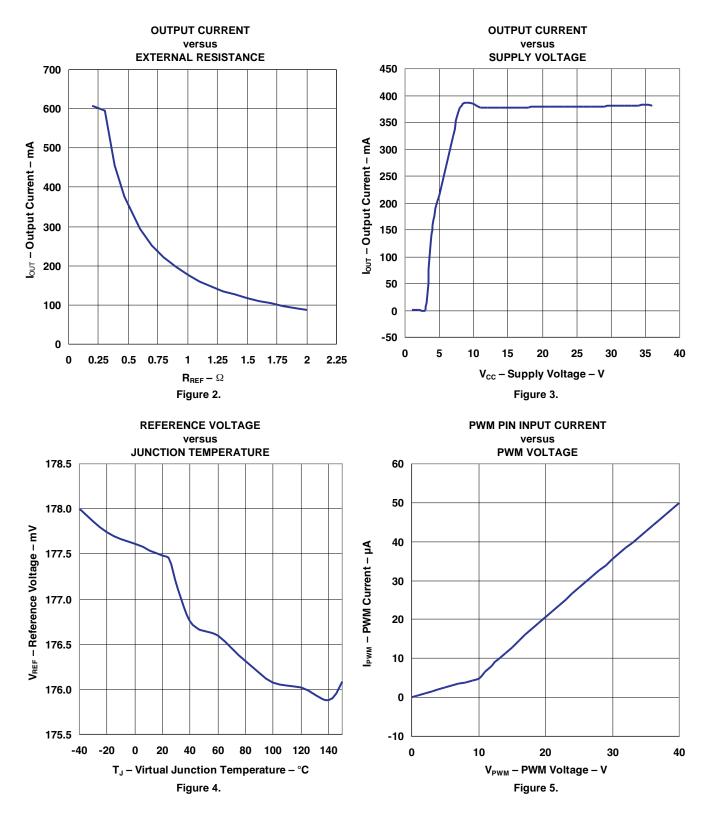
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{STHL}	Delay time, status reaction	$C_D = 47 \text{ nF}, \text{ ST H} \rightarrow \text{L}$	6	10	14	ms
t _{STLH}	Delay time, status release	$C_D = 47 \text{ nF}, \text{ ST } L \rightarrow H$		10	20	μs

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TYPICAL CHARACTERISTICS





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APPLICATION INFORMATION

Figure 6 shows a typical application with the TL4242-Q1 LED driver. A supply current adjusted by the R_{REF} resistor drives the three LEDs, preventing brightness variations due to forward voltage spread of the LEDs. An appropriate duty cycle applied to the PWM pin can compensate through software for the luminosity spread arising from the LED production process. Therefore, it is not necessary to select LEDs for forward voltage or luminosity classes. The minimum supply voltage calculates as the sum of the LED forward voltages, the TL4242-Q1 drop voltage (maximum 0.7 V at an LED current of 300 mA), and the maximum voltage drop at the shunt resistor R_{REF} of 185 mV.

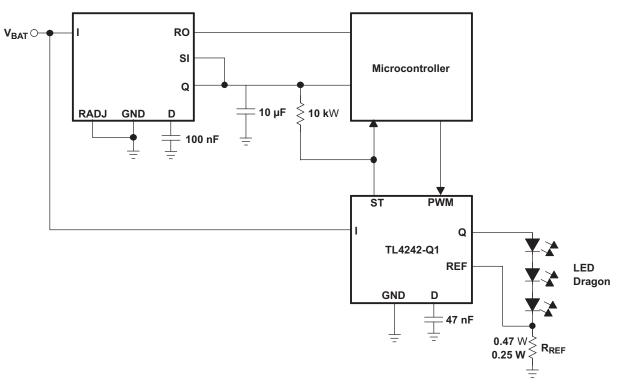


Figure 6. Application Circuit

The status output of the LED driver (ST) detects an open-load condition, enabling supervision of correct LED operation. A voltage drop at the shunt resistor (R_{REF}) below 25 mV (typical) detects an LED failure. In this case, the status output pin (ST) goes low after a delay time adjustable by an optional capacitor connected to pin D.

Figure 7 shows the functionality and timing of ST and PWM. One can adjust the status delay through the capacitor connected to pin D. Delay time scales linearly with the capacitance, C_D :

$$t_{\text{STHL,typ}} = \frac{C_{\text{D}}}{47 \text{ nF}} \times 10 \text{ ms}$$
$$t_{\text{STLH,typ}} = \frac{C_{\text{D}}}{47 \text{ nF}} \times 10 \text{ }\mu\text{s}$$

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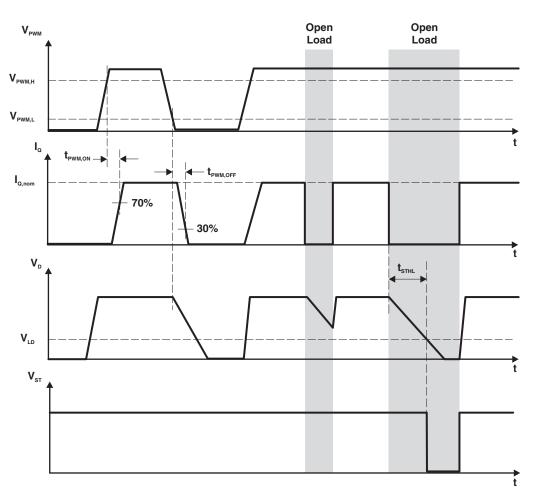


Figure 7. Function and Timing Diagram

Stoplight and Taillight Application

For many automobiles, the same set of LEDs illuminates both taillights and stoplights. Thus, the LEDs must operate at two different brightness levels, full brightness for the stoplight and 10% to 25% brightness for the taillight. The easiest way to achieve the different brightness is dimming by pulse-width modulation (PWM), which holds the color spectrum of the LED over its whole brightness range. The maximum current that passes through the LED is programmable by sense resistor R_{REF} .

Obtain the maximum current, I_{Qmax}, that passes through the LEDs by the following expression:

$$I_{Qmax} = \frac{V_{REF}}{R_{REF}}$$

For example, if R_{REF} equals 1 Ω , as V_{REF} is a fixed value range from 168 mV to 185 mV, I_{Qmax} should be 168 mA to 185 mA.



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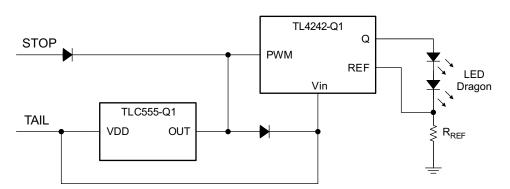


Figure 8. Stoplight and Taillight Application Circuit

Figure 8 shows the application circuit of the stoplight and taillight including an automotive-qualified timer, TLC555-Q1, the duty cycle of which is programmable by two external resistors. One can see that driving the STOP signal high pulls the PWM pin constantly high, creating 100% duty cycle. Thus the LEDs operate at full brightness. When the TAIL signal is high, the LEDs operate at 25% brightness because the TLC555-Q1 timer is programmed at a fixed duty cycle of 25%.

Thermal Information

This device operates a thermal shutdown (TSD) circuit as a protection from overheating. For continuous normal operation, the junction temperature should not exceed the thermal-shutdown trip point. If the junction temperature exceeds the thermal-shutdown trip point, the output turns off. When the junction temperature falls below the thermal-shutdown trip point, the output turns on again.

Calculate the power dissipated by the device according to the following formula:

$$\mathsf{P} = (\mathsf{V}_{\mathsf{I}} - \mathsf{V}_{\mathsf{O}}) \times \mathsf{I}_{\mathsf{O}} + \mathsf{V}_{\mathsf{I}} \times \mathsf{I}_{\mathsf{Q}}$$

In the formula, V_I represents the input voltage of the device, V_O stands for the output voltage, and I_O means the output current of LED and I_Q is the quiescent current dissipated by the device. The very small value of I_Q sometimes allows one to neglect it.

After determining the power dissipated by the device, calculate the junction temperature from the ambient temperature and the device thermal impedance.

$$\mathsf{T}_{\mathsf{J}} = \mathsf{T}_{\mathsf{A}} + \theta_{\mathsf{J}\mathsf{A}} \times \mathsf{P}$$

PCB Design Guideline

In order to prevent thermal shutdown, T_J must be less than 150°C. If the input voltage is very high, the power dissipation might be large. Currently there is the KTT (DDPAK) package which has good thermal impedance, but at the same time, the PCB layout is also very important. Good PCB design can optimize heat transfer, which is absolutely essential for the long-term reliability of the device.

- Maximize the copper coverage on the PCB to increase the thermal conductivity of the board, because the major heat-flow path from the package to the ambient is through the copper on the PCB. Maximjum copper is extremely important when there are not any heat sinks attached to the PCB on the other side of the package.
- Add as many thermal vias as possible directly under the package ground pad to optimize the thermal conductivity of the board.
- All thermal vias should be either plated shut or plugged and capped on both sides of the board to prevent solder voids. To ensure reliability and performance, the solder coverage should be at least 85 percent.

REVISION HISTORY

Changes from Revision B (September 2012) to Revision C Page Changes from Revision A (August, 2012) to Revision B Page

•	Removed package column in ordering information table.	. 2
•	Manually appended mechanical data, thermal pad data, and package option addendum	. 7

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TL4242QKTTRQ1	DDPAK/ TO-263	КТТ	7	500	330.0	24.4	10.6	15.8	4.9	16.0	24.0	Q2
TL4242TDRJRQ1	SON	DRJ	8	1000	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

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12-Oct-2012



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TL4242QKTTRQ1	DDPAK/TO-263	КТТ	7	500	340.0	340.0	38.0
TL4242TDRJRQ1	SON	DRJ	8	1000	210.0	185.0	35.0

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